## 1A, 1.5MHz Synchronous Step-Down Converter

## General Description

The AME5250A is a high efficiency monolithic synchronous buck regulator using a constant frequency, current mode architecture. Capable of delivering 1A output current over a wide input voltage range from 2.5 V to 5.5 V .

Supply current with no load is $400 \mu \mathrm{~A}$ and drops to $<1 \mu \mathrm{~A}$ in shutdown. The 2.5 V to 5.5 V input Voltage range makes the AME5250A ideally suited for single Li-lon batterypowered applications. 100\% duty cycle provides low dropout operation, extending battery life in portable systems. PWM pulse skipping mode operation provides very low output ripple voltage for noise sensitive applications. At very light load, the AME5250A will automatically skip pulses in pulse skip mode operation to maintain output regulation.

The internal synchronous switch increases efficiency and eliminates the need for an external Schottky diode. Low output voltages are easily supported with the 0.6 V feedback reference voltage. The AME5250A is available in small DFN-6D \& QFN-16C packages.

Other features include soft start, lower internal reference voltage with $2 \%$ accuracy, over temperature protection, and over current protection.

## Features

- High Efficiency: Up to $95 \%$
- Shutdown Mode Draws <1 $\mu \mathrm{A}$ Supply Current
- 2.5 V to 5.5 V Input Range
- Adjustable Output From 0.6 V to $\mathrm{V}_{\mathrm{IN}}$
- 1.0V, 1.2V, 1.5V, 1.6V, 1.8V, 2.5V and
3.3V Fixed/Adjustable Output Voltage
- 1A Output Current
- Low Dropout Operation: 100\% Duty Cycle
- No Schottky Diode Required
- 1.5 MHz Constant Frequency PWM Operation
- Small DFN-6D \& QFN-16C Packages
- All AME's Lead Free Product Meet RoHS

Standard

## Applications

- Cellular Telephones
- Personal Information Appliances
- Wireless and DSL Modems
- MP3 Players
- Portable Instruments


## Typical Application



Fixed Output Voltage
Figure 1. High Efficiency Step-Down Converter


## Adjustable Output Voltage

Figure 2. 1.8V at 1000 mA Step-Down Requlator

$$
\mathrm{C}_{\mathrm{Fw}}: 22 \mathrm{pF} \sim 220 \mathrm{pF}
$$

## 1A, 1.5MHz Synchronous Step-Down Converter

## ■ Function Block Diagram



Figure 3. Founction Block Diagram

Pin Configuration

DFN-6D
(2mmx2mmx0.75mm) Top View



## * Die Attach: <br> Conductive Epoxy

AME5250A-AWExxx


| 1. GND | 9. $I N$ |
| :--- | ---: |
| 2. GND | 10. IN |
| 3. GND | 11. IN |
| 4. FB/OUT | 12. IN |
| 5. GND | 13. SW |
| 6. NC | 14. SW |
| 7. EN | 15. SW |
| 8. NC | 16. NC |

1. GND
2. IN
3. NC
4. NC

* Die Attach:

Conductive Epoxy

Note:
The area enclosed by dashed line represents Exposed Pad and connect to GND.

## Pin Description

| Pin Number |  | Pin Name | Pin Description |
| :---: | :---: | :---: | :--- |
| DFN | QFN |  | NC |
| 1 | $6,8,16$ | No connection. Not internally connected. Can left floating or <br> connected to GND. |  |
| 2 | 7 | EN | Enable Control Input, active high. |
| 3 | $9,10,11,12$ | IN | Input Supply Voltage Pin. <br> Bypass this pin with a capacitor as close to the device as <br> possible. |
| 4 | $13,14,15$ | SW | Switch Node Connection to Inductor. |
| 5 | $1,2,3,5$ | GND | Ground. Tie directly to ground plane. |
| 6 | 4 | FB/OUT | Output voltage Feedback input. |

■ Ordering Information


| Pin Configuration \& Special Feature | Package Type | Number of Pins | Output Voltage |
| :---: | :---: | :---: | :---: |
|  | V: DFN <br> W: QFN | $\begin{aligned} & Y: 6 \\ & \text { E: } 16 \end{aligned}$ | $\begin{aligned} & \text { 100: } 1.0 \mathrm{~V} \\ & \text { 120: } 1.2 \mathrm{~V} \\ & \text { 150: } 1.5 \mathrm{~V} \\ & \text { 160: } 1.6 \mathrm{~V} \\ & \text { 180: } 1.8 \mathrm{~V} \\ & \text { 250: } 2.5 \mathrm{~V} \\ & \text { 330: } 3.3 \mathrm{~V} \\ & \text { ADJ: Adjustable } \end{aligned}$ |

Absolute Maximum Ratings

| Parameter | Symbol | Maximum | Unit |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Input Supply Voltage | $\mathrm{V}_{\text {IN }}$ | -0.3 to 6.5 | V |  |  |
| EN, $\mathrm{V}_{\text {OUT }}$ Voltage | $\mathrm{V}_{\text {EN }}, \mathrm{V}_{\mathrm{OUT}}$ | -0.3 to $\mathrm{V}_{\mathrm{IN}}$ |  |  |  |
| SW Voltage | $\mathrm{V}_{\text {SW }}$ | -0.3 to $\mathrm{V}_{\mathrm{IN}}$ |  |  |  |
| ESD Classification | $\mathrm{B}^{*}$ |  |  |  |  |

Caution: Stress above the listed absolute maximum rating may cause permanent damage to the device.

* HBM B: 2000V~3999V


## ■ Recommended Operating Conditions

| Parameter | Symbol | Rating | Unit |
| :--- | :---: | :---: | :---: |
| Supply Voltage Voltage | $\mathrm{V}_{\mathrm{IN}}$ | 2.5 to 5.5 | V |
| Ambient Temperature Range | $\mathrm{T}_{\mathrm{A}}$ | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Junction Temperature Range | $\mathrm{T}_{\mathrm{J}}$ | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |

■ Thermal Information

| Parameter | Package | Die Attach | Symbol | Maximum | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Thermal Resistance* (Junction to Case) | DFN-6D | Conductive Epoxy | $\theta_{\text {Jc }}$ | 85 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Thermal Resistance (Junction to Ambient) |  |  | $\theta_{\text {JA }}$ | 160 |  |
| Internal Power Dissipation |  |  | $\mathrm{P}_{\mathrm{D}}$ | 625 | mW |
| Thermal Resistance* (Junction to Case) | QFN-16C | Conductive Epoxy | $\theta_{\mathrm{Jc}}$ | 67 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Thermal Resistance (Junction to Ambient) |  |  | $\theta_{\mathrm{JA}}$ | 149 |  |
| Internal Power Dissipation |  |  | $\mathrm{P}_{\mathrm{D}}$ | 670 | mW |
| Solder Iron (10Sec)** |  |  |  | 350 | ${ }^{\circ} \mathrm{C}$ |

* Measure $\theta_{\mathrm{Jc}}$ on backside center of Exposed Pad.
** MIL-STD-202G 210F


## - Electrical Specifications

$\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{OUT}}=2.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{FB}}=0.6 \mathrm{~V}, \mathrm{~L}=2.2 \mu \mathrm{H}, \mathrm{C}_{\mathrm{IN}}=4.7 \mu \mathrm{~F}, \mathrm{C}_{\mathrm{OUT}}=10 \mu \mathrm{~F}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{I}_{\mathrm{MAX}}=1 \mathrm{~A}$ unless otherwise specified.

| Parameter | Symbol | Test Condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage | $\mathrm{V}_{\text {IN }}$ |  | 2.5 |  | 5.5 | V |
| Output Voltage Accuracy | $\Delta \mathrm{V}_{\text {OUT }}$ | $\mathrm{V}_{\text {IN }}=2.5$ to 5.5 V , in PWM mode For Fixed Output Voltage | -3 |  | 3 | \% |
| Adjustable Output Range | $\mathrm{V}_{\text {out }}$ |  | $\mathrm{V}_{\mathrm{FB}}$ |  | $\mathrm{V}_{\mathrm{IN}^{-}-0.2}$ | V |
| Feedback Voltage | $V_{\text {FB }}$ | For Adjustable OutputVoltage | 0.588 | 0.6 | 0.612 | V |
| Feedback Pin Bias Current | $\mathrm{I}_{\text {FB }}$ | $\mathrm{V}_{\mathrm{FB}}=\mathrm{V}_{\mathrm{IN}}$ | -50 |  | 50 | nA |
| Quiescent Current | $\mathrm{I}_{\mathrm{Q}}$ | $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}, \mathrm{~V}_{\mathrm{FB}}=1 \mathrm{~V}$ |  | 0.4 | 0.5 | mA |
| Shutdown Current | $\mathrm{I}_{\text {SHDN }}$ | $\mathrm{V}_{\mathrm{EN}}=\mathrm{GND}$ |  | 0.1 | 1 | $\mu \mathrm{A}$ |
| Switch Frequency | $\mathrm{f}_{\text {OSC }}$ |  | 1.2 | 1.5 | 1.8 | MHz |
| High-side Switch On-Resistance | $\mathrm{R}_{\mathrm{DS}, \mathrm{ON}, \mathrm{LHI}}$ | $\mathrm{I}_{\mathrm{SW}}=200 \mathrm{~mA}, \mathrm{~V}_{\mathrm{IN}}=3.6 \mathrm{~V}$ |  | 0.28 |  | $\Omega$ |
| Low-side Switch On-Resistance | $\mathrm{R}_{\text {DS,ON, LO }}$ | $\mathrm{I}_{\mathrm{SW}}=200 \mathrm{~mA}, \mathrm{~V}_{\text {IN }}=3.6 \mathrm{~V}$ |  | 0.25 |  | $\Omega$ |
| Switch Current Limit | $\mathrm{I}_{\text {SW,CL }}$ | $\mathrm{V}_{1 \mathrm{~N}}=2.5$ to 5.5 V | 1.4 | 1.6 |  | A |
| EN High (Enabled the Device) | $\mathrm{V}_{\text {EN,HI }}$ | $\mathrm{V}_{\text {IN }}=2.5$ to 5.5 V | 1.5 |  |  | V |
| EN Low (Shutdown the Device) | $\mathrm{V}_{\text {EN,LO }}$ | $\mathrm{V}_{\text {IN }}=2.5$ to 5.5 V |  |  | 0.4 | V |
| Input Undervoltage Lockout | $\mathrm{V}_{\text {UVLO }}$ | rising edge |  | 1.8 |  | V |
| Input Undervoltage Lockout Hysteresis | $\mathrm{V}_{\text {UVLO,HYST }}$ |  |  | 0.1 |  | V |
| Thermal Shutdown Temperature | OTP | Shutdown, temperature increasing |  | 160 |  | ${ }^{\circ} \mathrm{C}$ |
| Maximum Duty Cycle | $\mathrm{D}_{\text {MAX }}$ |  | 100 |  |  | \% |
| SW Leakage Current |  | $\begin{gathered} \mathrm{EN}=0 \mathrm{~V}, \mathrm{~V}_{\mathbb{I N}}=5.0 \mathrm{~V} \\ \mathrm{~V}_{\mathrm{SW}}=0 \mathrm{~V} \text { or } 5.0 \mathrm{~V} \end{gathered}$ | -1 |  | 1 | $\mu \mathrm{A}$ |

## AME5250A

## 1A, 1.5MHz Synchronous Step-Down Converter

## Detailed Description

## Main Control Loop

The AME5250A uses a constant frequency, current mode step-down architecture. Both the main (P-channel MOSFET) and synchronous ( N -channel MOSFET) switches are intermal. During normal operation, the internal top power MOSFET is turned on each cycle when the oscillator sets the RS latch, and turned off when the current comparator resets the RS latch. While the top MOSFET is off, the bottom MOSFET is turned on until either the inductor current starts to reverse as indicated by the current reversal comparator IRCMP.

## Pulse Skipping Mode Operation

At light loads, the inductor current may reach zero or reverse on each pulse.The bottom MOSFET is turned off by the current reversal comparator, IRCMP, and the switch voltage will ring. This is discontinuous mode operation, and is normal behavior for the switching regulator.

## Short-Circuit Protection

When the output is shorted to ground, the frequency of the oscillator is reduced to about 180 KHz . This frequency foldback ensures that the inductor current hsa more time do decay, thereby preventing runaway. The oscillator's frequency will progressively increase to 1.5 MHz when $\mathrm{V}_{\mathrm{FB}}$ or $\mathrm{V}_{\mathrm{OUT}}$ rises above 0 V .

## Dropout Operation

As the input supply voltage decreases to a value approaching the output voltage, the duty cycle increases toward the maximum on-time. Further reduction of the supply voltage forces the main switch to remain on for more than one cycle until it reaches $100 \%$ duty cycle. The output voltage will then be determined by the input voltage minus the voltage drop across the P-channel MOSFET and the inductor.

## Application Information

The basic AME5250A application circuit is shown in Typical Application Circuit. External component selection is determined by the maximum load current and begins with the selection of the inductor value and followed by $\mathrm{C}_{\mathrm{N}}$ and $\mathrm{C}_{\text {out }}$.

## Inductor Selection

For a given input and output voltage, the inductor value and operating frequency determine the ripple current. The ripple current DIL increases with higher $\mathrm{V}_{\mathbb{N}}$ and decreases with higher inductance.

$$
\Delta I_{L}=\frac{1}{f \times L} \times V_{\text {OUT }}\left(1-\frac{V_{\text {OUT }}}{V_{I N}}\right)
$$

A reasonable starting point for setting ripple current is $\Delta \mathrm{I}_{\mathrm{L}}=0.4$ (Imax). The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation. For better efficiency, choose a low DC-resistance inductor.

## $\mathrm{C}_{\text {IN }}$ and $\mathrm{C}_{\text {out }}$ Selection

The Input capacitance, $\mathrm{C}_{\mathbb{N}}$ is needed to filter the trapezoidal current at the source of the top MOSFET. To prevent large voltage transients, a low ESR input capacitorsized for the maximum RMS current must be used. The maximum RMS capacitor current is given by:

$$
I_{\text {RMS }}=I_{\text {OUT (MAX) }} \times \frac{V_{\text {OUT }}}{V_{I N}} \times \sqrt{\frac{V_{I N}}{V_{\text {OUT }}}-1}
$$

This formula has a maximum at $\mathrm{V}_{\mathrm{IN}}=2 \mathrm{~V}_{\text {OUT }}$, where $\mathrm{I}_{\text {RMS }}=\mathrm{I}_{\text {out }} / 2$. This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief. Note that the capacitor manufacturer ripple current ratings are often based on 2000 hours of life. This makes it advisable to further derate the capacitor, or choose a capacitor rated at a higher temperature than required.

## AME5250A

## 1A, 1.5MHz Synchronous Step-Down Converter

The selection of $\mathrm{C}_{\text {out }}$ is determined by the effective series resistance(ESR) that is required to minimize voltage ripple and load step transients. The output ripple, $\mathrm{V}_{\text {out }}$, is determined by:

$$
\Delta V_{\text {OUT }} \cong \Delta I_{L}\left(E S R+\frac{1}{8 f C_{\text {OUT }}}\right)
$$

## Using Ceramic Input and Output Capacitors

Higher values, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. However, care must be taken when these capacitors are used at the input and output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input, $\mathrm{V}_{\mathbb{I N}}$. At best, this ringing can couple to the output and be mistaken as loop instability. At worst, a sudden inrush of current through the long wires can potentially cause a voltage spike at $\mathrm{V}_{\text {IN }}$ large enough to damage the part.

## Output Voltage Programming

The output voltage is set by an external resistive divider according to the following equation:

$$
V_{\text {OUT }}=V_{R E F} \times\left(1+\frac{R 1}{R 2}\right)
$$

Where $\mathrm{V}_{\text {REF }}$ equals to 0.6 V typical. The resistive divider allows the FB pin to sense a fraction of the output voltage as shown in Figure 4.


Figure 4. Setting the AME5250A Output Voltage


Figure 5. 1.2V Step-Down Regulator
$\mathrm{C}_{\mathrm{FW}}$ : 22pF~220pF


Figure 6. 1.5V Step-Down Regulator

$$
\mathrm{C}_{\mathrm{Fw}}: 22 \mathrm{pF} \sim 220 \mathrm{pF}
$$



Figure 7.2.5V Step-Down Regulator
$\mathrm{C}_{\mathrm{Fw}}$ : 22pF~220pF


Figure 8. 1.6V Step-Down Regulator

$$
\mathrm{C}_{\mathrm{Fw}}: 22 \mathrm{pF} \sim 220 \mathrm{pF}
$$



Figure 9.3.3V Step-Down Regulator
$\mathrm{C}_{\mathrm{Fw}}$ : 22pF~220pF

## 1A, 1.5MHz Synchronous Step-Down Converter

## PC Board Layout Checklist

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the AME5250A. These items are also illustrated graphically in Figures 10 and Figures 11. Check the following in your layout:

1. The power traces, consisting of the GND trace, the SW trace and the $\mathrm{V}_{\text {IN }}$ trace should be kept short, direct and wide.
2. Does the $\mathrm{V}_{\text {FB }}$ pin connect directly to the feedback resistors? The resistive divider R2/R1 must be connected between the (+) plate of $\mathrm{C}_{\text {out }}$ and ground.
3. Does the (+) plate of $\mathrm{C}_{\text {IN }}$ connect to $\mathrm{V}_{\text {IN }}$ as closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
4. Keep the switching node, SW , away from the sensitive $\mathrm{V}_{\mathrm{FB}}$ node.
5. Keep the (-) plates of $\mathrm{C}_{\mathbb{N}}$ and $\mathrm{C}_{\text {out }}$ as close as possible.


Figure 10. AME5250AAdjustable Voltage
Regulator Layout Diagram


Reglator


Figure 11. AME5250A Fixed Voltage Regulator Layout Diagram

## 1A, 1.5MHz Synchronous Step-Down Converter

## Application Information

External components selection

| Supplier | Inductance <br> $(\mu \mathrm{H})$ | Current Rating <br> $(\mathbf{m A})$ | DCR <br> $(\mathbf{m} \Omega)$ | Dimensions <br> $(\mathrm{mm})$ | Series |
| :--- | :---: | :---: | :---: | :---: | :--- |
| TAIYO YUDEN | 2.2 | 1480 | 60 | $3.00 \times 3.00 \times 1.50$ | NR 3015 |
| GOTREND | 2.2 | 1500 | 58 | $3.85 \times 3.85 \times 1.80$ | GTSD32 |
| Sumida | 2.2 | 1500 | 75 | $4.50 \times 3.20 \times 1.55$ | CDRH2D14 |
| Sumida | 4.7 | 1000 | 135 | $4.50 \times 3.20 \times 1.55$ | CDRH2D14 |
| TAIYO YUDEN | 4.7 | 1020 | 120 | $3.00 \times 3.00 \times 1.50$ | NR 3015 |
| GOTREND | 4.7 | 1100 | 146 | $3.85 \times 3.85 \times 1.80$ | GTSD32 |

Table 1. Recommended Inductors

| Supplier | Capacitance <br> $(\mu \mathrm{H})$ | Package | Part Number |
| :--- | :---: | :---: | :--- |
| TDK | 4.7 | 603 | C1608JB0J475M |
| MURATA | 4.7 | 603 | GRM188R60J475KE19 |
| TAIYO YUDEN | 4.7 | 603 | JMK107BJ475RA |
| TAIYO YUDEN | 10 | 603 | JMK107BJ106MA |
| TDK | 10 | 805 | C2012JB0J106M |
| MURATA | 10 | 805 | GRM219R60J106ME19 |
| MURATA | 10 | 805 | GRM219R60J106KE19 |
| TAIYO YUDEN | 10 | 805 | JMK212BJ106RD |

Table 2. Recommended Capacitors for $\mathrm{C}_{\mathrm{IN}}$ and $\mathrm{C}_{\text {OUT }}$

## 1A, 1.5MHz Synchronous Step-Down Converter

Characterization Curve
Efficiency vs. Output Current


Efficiency vs. Output Current



## - Characterization Curve



Frequency vs. Supply Voltage


Current Limit vs. Temperature


Frequency vs. Temperature


Output Voltage vs. Output Current


Current Limit vs. Temperature


AME
AME5250A

## 1A, 1.5MHz Synchronous Step-Down Converter

## Characterization Curve

Current Limit vs. Temperature


Light Load Mode Output Voltage Ripple

$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}$
$\mathrm{~V}_{\text {OUT }}=1.8 \mathrm{~V}$
$\mathrm{I}_{\text {OUT }}=50 \mathrm{~mA}$

1) $V_{s w}=2 V /$ div
2) $V_{\text {Out }}=10 \mathrm{mV} / \mathrm{Div}$
3) $I_{\llcorner }=500 \mathrm{~mA} / \mathrm{Div}$

Heavy Load Mode Output Voltage Ripple

$\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}$
$V_{\text {OUT }}=1.2 \mathrm{~V}$
$\mathrm{I}_{\text {OUT }}=1 \mathrm{~A}$

1) $V_{s w}=2 V / d i v$
2) $V_{\text {out }}=10 \mathrm{mV} /$ Div
3) $I_{L}=500 \mathrm{~mA} / \mathrm{Div}$

Load Step

$40 \mu \mathrm{~S} / \mathrm{Div}$
$\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}$
$\mathrm{V}_{\text {OUT }}=1.8 \mathrm{~V}$
$\mathrm{I}_{\text {OUT }}=0 \mathrm{~A} \sim 1 \mathrm{~A} \sim 0 \mathrm{~A}$

1) $\mathrm{V}_{\text {OUT }}=100 \mathrm{mV} /$ Div
2) $I_{\text {out }}=500 \mathrm{~mA} /$ Div

## ■ Characterization Curve

Load Step

$40 \mu \mathrm{~S} /$ Div
$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}$
$\mathrm{V}_{\text {OUT }}=1.8 \mathrm{~V}$
$\mathrm{I}_{\text {OUT }}=50 \mathrm{~mA} \sim 1 \mathrm{~A} \sim 50 \mathrm{~mA}$

1) $V_{\text {OUT }}=100 \mathrm{mV} / \mathrm{Div}$
2) $I_{\text {out }}=500 \mathrm{~mA} /$ Div

Power On from EN

$400 \mu \mathrm{~S} / \mathrm{Div}$

$$
\begin{aligned}
& \mathrm{V}_{\text {out }}=1.2 \mathrm{~V} \\
& \mathrm{I}_{\text {out }}=1 \mathrm{~A}
\end{aligned}
$$

1) $\mathrm{EN}=2 \mathrm{~V} / \mathrm{Div}$
2) $V_{\text {out }}=500 \mathrm{mV} / \mathrm{Div}$
3) $I_{L}=1 \mathrm{~A} / \mathrm{Div}$

Load Step

$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}$
$\mathrm{V}_{\text {OUT }}=1.8 \mathrm{~V}$
$\mathrm{I}_{\text {out }}=200 \mathrm{~mA} \sim 1 \mathrm{~A} \sim 200 \mathrm{~mA}$

1) $V_{\text {OUT }}=100 \mathrm{mV} /$ Div
2) $I_{\text {OUT }}=500 \mathrm{~mA} / \mathrm{div}$

Power Off from EN

$50 \mu \mathrm{~S} / \mathrm{Div}$
$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}$
$\mathrm{V}_{\text {out }}^{\text {iv }}=1.8 \mathrm{~V}$
$\mathrm{l}_{\text {OUT }}=1 \mathrm{~A}$

1) $E N=2 V / D i v$
2) $V_{\text {out }}=2 \mathrm{~V} / \mathrm{Ddiv}$
3) $I_{L}=500 \mathrm{~mA} / \mathrm{Div}$

## 1A, 1.5MHz Synchronous Step-Down Converter

## ■ Tape and Reel Dimension

DFN-6D
(2mmx2mmx0.75mm)


Carrier Tape, Number of Components Per Reel and Reel Size

| Package | Carrier Width (W) | Pitch (P) | Part Per Full Reel | Reel Size |
| :---: | :---: | :---: | :---: | :---: |
| DFN-6D <br> $(2 \times 2 \times 0.75 \mathrm{~mm})$ | $8.0 \pm 0.1 \mathrm{~mm}$ | $4.0 \pm 0.1 \mathrm{~mm}$ | 3000 pcs | $180 \pm 1 \mathrm{~mm}$ |

QFN-16C
(3mmx3mmx0.75mm)


Carrier Tape, Number of Components Per Reel and Reel Size

| Package | Carrier Width (W) | Pitch (P) | Part Per Full Reel | Reel Size |
| :---: | :---: | :---: | :---: | :---: |
| QFN-16C <br> $(3 \times 3 \times 0.75 \mathrm{~mm})$ | $12.0 \pm 0.1 \mathrm{~mm}$ | $4.0 \pm 0.1 \mathrm{~mm}$ | 3000 pcs | $330 \pm 1 \mathrm{~mm}$ |

## Package Dimension

DFN-6D
(2mmx2mmx0.75mm)


REAR VIEW

| SYMBOLS | MILLIMETERS |  | INCHES |  |
| :---: | :---: | :---: | :---: | :---: |
|  | MIN | MAX | MIN | MAX |
| A | 0.700 | 0.800 | 0.028 | 0.031 |
| D | 1.900 | 2.100 | 0.075 | 0.083 |
| E | 1.900 | 2.100 | 0.075 | 0.083 |
| e | 0.650 TYP |  | 0.026 TYP |  |
| D1 | 1.100 | 1.650 | 0.043 | 0.065 |
| E1 | 0.600 | 1.050 | 0.024 | 0.041 |
| b | 0.180 | 0.350 | 0.007 | 0.014 |
| L | 0.200 | 0.450 | 0.008 | 0.018 |
| G | 0.178 | 0.228 | 0.007 | 0.009 |
| G1 | 0.000 | 0.050 | 0.000 | 0.002 |

## 1A, 1.5MHz Synchronous Step-Down Converter

Package Dimension
QFN-16C
(3mmx3mmx0.75mm)


Top View


| SYMBOLS | MILLIMETERS |  | INCHES |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN | MAX | MIN | MAX |  |  |
| A | 0.700 | 0.800 | 0.028 | 0.031 |  |  |
| A1 | 0.000 | 0.050 | 0.000 | 0.002 |  |  |
| A3 | $0.203 R$ REF. |  | $0.008 R$ REF. |  |  |  |
| D | 2.924 | 3.076 | 0.115 | 0.121 |  |  |
| E | 2.924 | 3.076 | 0.115 | 0.121 |  |  |
| D1 | 1.600 | 1.800 | 0.063 | 0.071 |  |  |
| E1 | 1.600 | 1.800 | 0.063 | 0.071 |  |  |
| k | 0.200 MIN.$$ |  | 0.008 MIN.$$ |  |  |  |
| b | 0.180 | 0.280 | 0.007 |  |  |  |
| e | 0.500 TYP. |  | 0.020 TYP. |  |  |  |
| L | 0.324 |  | 0.476 | 0.013 |  | 0.019 |

Lead pattern
DFN-6D
(2mmx2mmx0.75mm)


Note:

1. Lead pattern unit description:

BSC: Basic. Represents theoretical exact dimension or dimension target.
2. Dimensions in Millimeters.
3. General tolerance $\pm 0.05 \mathrm{~mm}$ unless otherwise specified.

Lead pattern
QFN-16C
(3mmx3mmx0.75mm)


Note:

1. Lead pattern unit description:

BSC: Basic. Represents theoretical exact dimension or dimension target.
2. Dimensions in Millimeters.
3. General tolerance $\pm 0.05 \mathrm{~mm}$ unless otherwise specified.

## Life Support Policy:

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AME, Inc. reserves the right to make changes in the circuitry and specifications of its devices and advises its customers to obtain the latest version of relevant information.
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